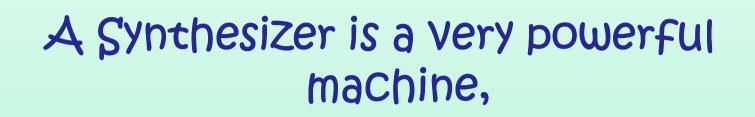
# MODERN FREQUENCY SYNTHESIZERS

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**RSGB** Convention 2014



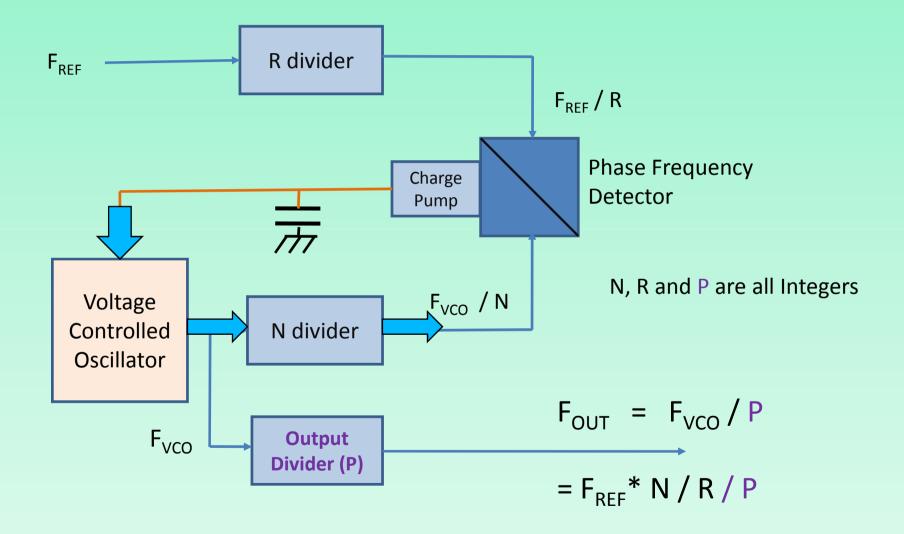
It can generate all frequencies once,

And most of them twice,

Simultaneously.

(G3RZP, sometime in the 1980's)

### The Basic Integer- N Synth



- Tuning Grid in steps of F<sub>COMP</sub> / P
- Loop time constant / bandwidth, spurii all depend on F<sub>COMP</sub> which is defined by the Step Size
- Useful for
  - Fixed sources eg. transverter LOs
  - Channelised operation
  - Multiplier from a variable input , eg. DDS, VFO
- F<sub>OUT</sub> = F<sub>IN</sub> \* N / R / P : Non-integer Freq multiplier

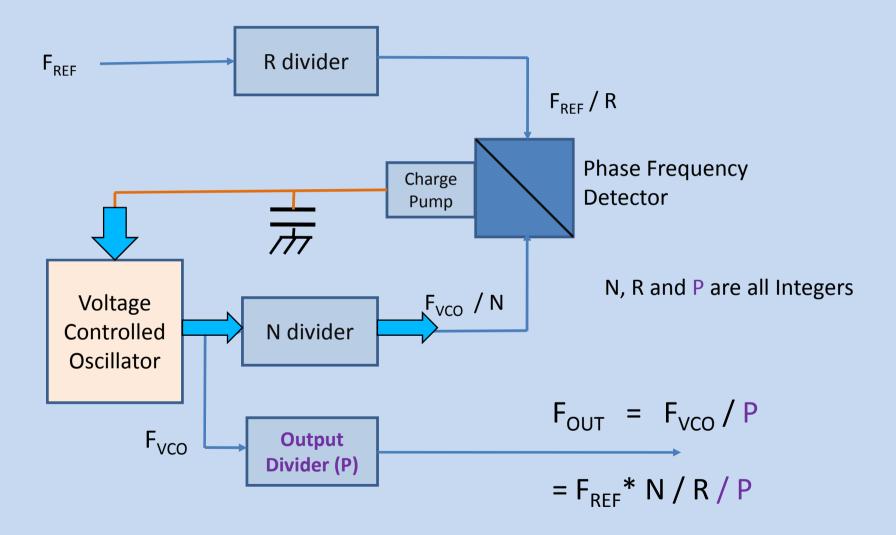
### Phase Noise

- <u>The Performance Limiter</u>
- Generated as the loop continually adjusts itself
  - Several Sources
- Noise appears as FM / phase jitter on the carrier.
  - INSIDE the Loop Bandwidth it is determined by the REFERENCE alone
  - OUTSIDE the Loop Bandwidth it is determined mainly by the VCO and its drive circuitry.

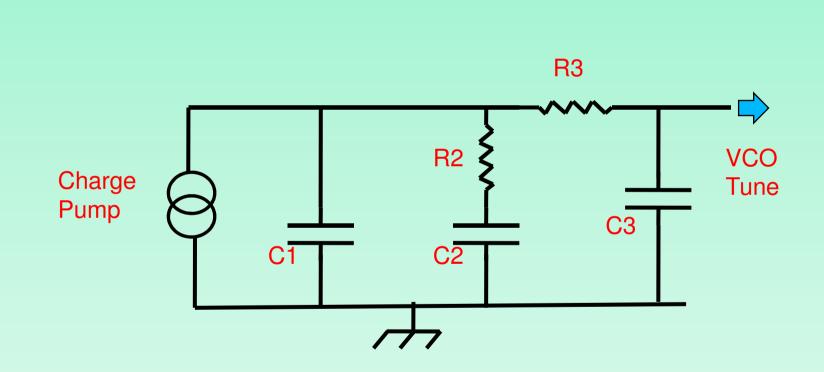
### Loop Bandwidth

- The Loop Filter defines the loop performance
  - Ideally set at the point where Reference and VCO noise cross over ...... BUT ......
  - Has to attenuate comparison frequency sidebands
  - Prevent External influences (microphony)
  - Determines Lockup Speed several cycles
- Often want as high a loop BW as possible,
  - Subject to loop components themselves not contributing.
  - Filter Order, good design techniques allow 4<sup>th</sup> order filters
- Programmable Charge Pump allows post design tweaking

#### The Basic Integer- N Synth





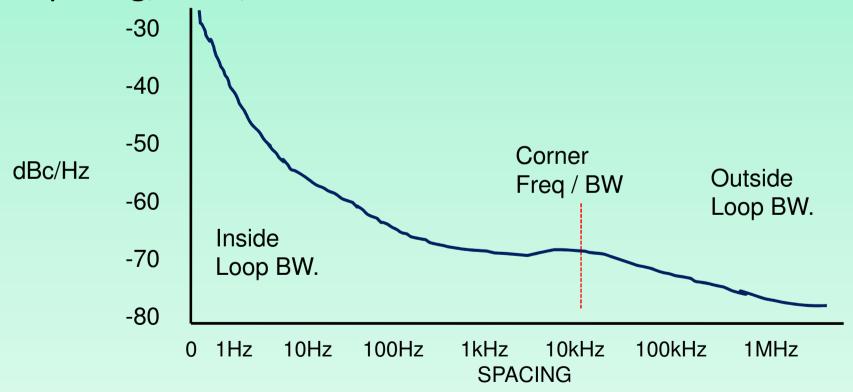


### Phase Noise continued

- Far-out phase noise is mostly a strong / adjacent signal problem. You need to decide how important that is:
  - On Transmit: spurii, cause QRM, regulatory
  - On Receive: Strong signals, reciprocal mixing
- Close in, Phase Noise affects audibility, sounds like rumble.
  - Use a low jitter good quality reference
  - Rarely a problem with latest generation synth chips
- Output Divider reduces Phase Noise by N<sup>2</sup>
  - So a GHz synthesizer divided down can give a good HF or VHF source

# dBC/Hz

Relative amplitude of the phase noise, in a 1Hz bandwidth referenced to the carrier. Specified at a particular spacing, in Hz, from that carrier



### Latest Advances in Modern Devices

- Phase Frequency detectors work to many MHz
  - High F<sub>COMP</sub> means high loop bandwidth
  - Improves close in phase noise
  - Programmable charge pump
- Integrated output divider
  - Gives UHF to low VHF coverage and smaller steps
- Integrated VCO on the chip
  - (GHz only, but *can* be octave coverage)

### Internal VCO

- One chip Synthesizer
  - No Need for separate VCO hardware / module
    - Packaged VCOs often more expensive than the synth chip itself
  - No High tuning voltages
    - Many packaged VCOs need 20V
- BUT
  - Higher far-out Phase Noise due to lower Q silicon resonator (outside loop bandwidth)
  - Need more programming / setting up

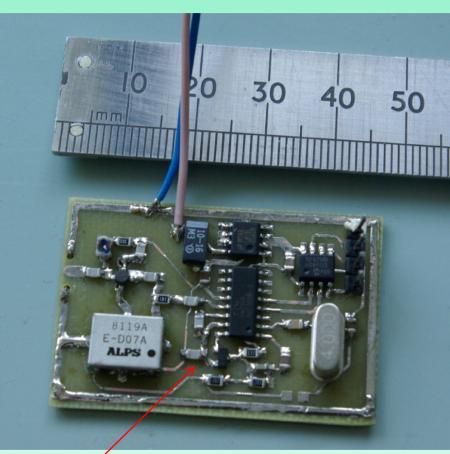
### Internal VCO

- Wide(ish) tuning range,
  - Needs several chip variants
  - Removes Tuning voltage limitation
- Auto calibration pretunes the VCO
  - Typical sensitivity 20 40MHz/V in spite of 100's of MHz tuning range
  - The chip has to be "told" the value of  $F_{REF}$
  - Auto Recal <u>whether you like it or not</u> = glitches when frequency is changed (sometimes)

## A Few useful Integer-N chips

- LMX 23xx, 24xx (many family members)
  - Obsolete, but the workhorse of much surplus stuff
  - 100MHz to 2.8GHz
  - $F_{COMP}$  to 1MHz
- ADF4110/3 is pin compatible with LMX2326, and goes to 4GHz
- LTC6946 Internal VCO,
  - 3 overlapping band variants 2.2 5.8GHz
  - Output divider, 1 6, so goes down to 370MHz
  - F<sub>COMP</sub> to 100MHz
    - There are lots of others out there.

#### Avoid Sat-TV Type Chips such as SP6289



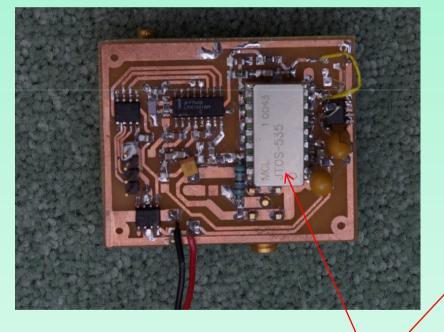
External NPN Charge Pump Amplifier allows high tuning voltage (30V)

Low F<sub>COMP</sub> - fixed prescalar Typical 8kHz Low Loop Bandwidth (100Hz) Large Step size (125kHz) Poor Phase Noise

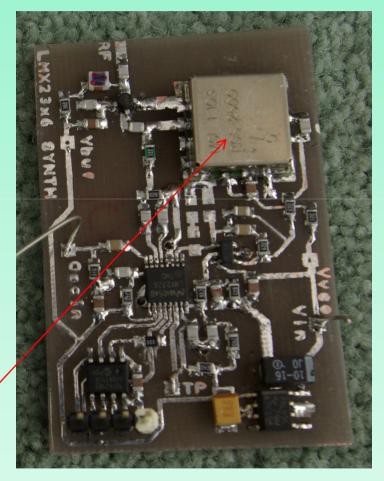
#### Totally and utterly useless for narrowband designs

#### Worth using if you can find them

LMX1501 Very Obsolete, but OK for V/UHF. Simple programming



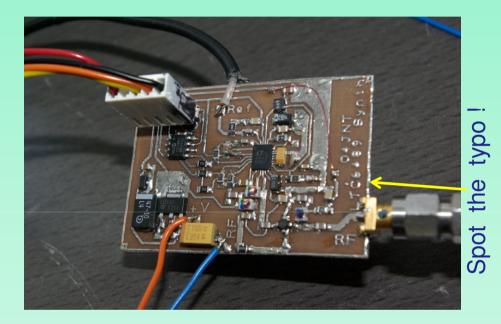
#### LMX2326

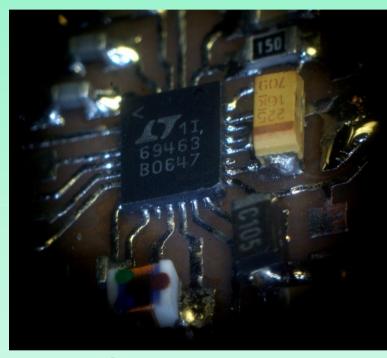


Packaged VCOs

There are many others

#### LTC6946, integrated 3-6GHz VCO





28 pin LQFP package Home built PCB, chip installed using the solder-wipe technique with liquid flux

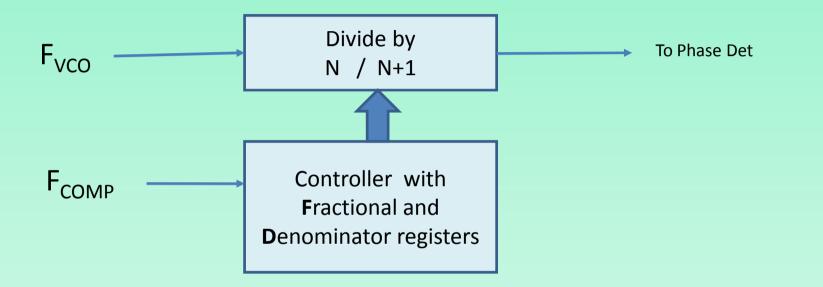
Used by GW4DGU in 10GHz Transverter LO

BUT Integer-N designs still only offer a coarse frequency setting grid.  $F_{COMP} / P$ 

## The Next stage - Tiny steps

- A low  $F_{\rm COMP}$  is not feasible if we want to keep loop bandwidth as high as possible
  - 10Hz steps would need 1Hz bandwidth 20s lockup!
- Generate high frequencies and divide down limited use
  - Although for LF, dividing by 1000 10000 works.
- Divide down and mix,
  - Complicated, spurii, hardware filtering
  - Yesteryear's solution
  - Many Commercial & amateur radios used triple loop synthesizers for 10 or 1Hz steps.
- DDS + PLL multiplier: Works, but spurii can be bad.
- So Use ideas learned from DSP, Fast processing, filter out the crud !

#### The Fractional-N Synth



Periodically change the value of N to N+1, F times out of every D, N' = N + F / D  $F_{OUT} = F_{COMP} * (N + F / D)$ Fractional counter runs at  $F_{COMP}$  which can be many MHz Tuning grid =  $F_{COMP} / D$ . Large D = small tuning steps

## BUT !

#### With this simple route There will be multiple SPURII at the grid spacing.

They can be controlled. Their amplitude can be made low *enough*. They can be made to just look like phase noise. Usually On a good day

### 'Real' Fractional-N Synthesizers

- Use N / N+1 / N+2 / N+3 Called "Sigma-Delta "
  - Shift between them in a pseudo random way
  - Dither to spread out the spurii.
  - Do clever stuff
- All this happens inside the chip
- As well as other things the manufacturers don't want to tell us!
- They let the users programme everything, optimising by trial and error. *Is this a cop-out?*

### The Result

- A 22 Bit Denominator (4194304) gives fine steps even with R = 1, where  $F_{COMP} =$  Input Reference
  - 10MHz in
  - means step size can be 2.4Hz
  - We can define specific step sizes for Multi-Frequency data modes. Eg 5.3833Hz for JT65B (D = 1857596)
- High F<sub>COMP</sub> means Fract-N jitter is kept low
- Loop bandwidth stays high. 20 100kHz typ.
- THE IDEAL SOLUTION TO EVERYTHING ?
  - Almost ----- but not quite!

### Limitations

• Problems around low and max values of F,

– Ie. Close to integer values

- Manufacturers provide a host of user settable options.
  - Usually possible to find one that works
  - Unless you can't

#### Just a few of the Synth Chips around

#### Integer-N

Type No	VCO	Freq range	O/P Divider	Vcc [max] (Vph)	Availability
LMX2326	External	100 – 2800	_	5 (5)	Obsolete, found on Surplus PCBs
ADF4110 / 3		80 - 550 / 4000	-	5 (5)	RS $\pm 4.50$ Pinout = LMX2326
, ADF4118	External	100 – 3000	-	5 (5)	RS £3.20 Pinout = LMX2326
LTC6946	Internal (3)	2240 – 5790	1-6	3.3 & 5 dual	£12 ++ US Stock Farnell

#### **Fractional-N**

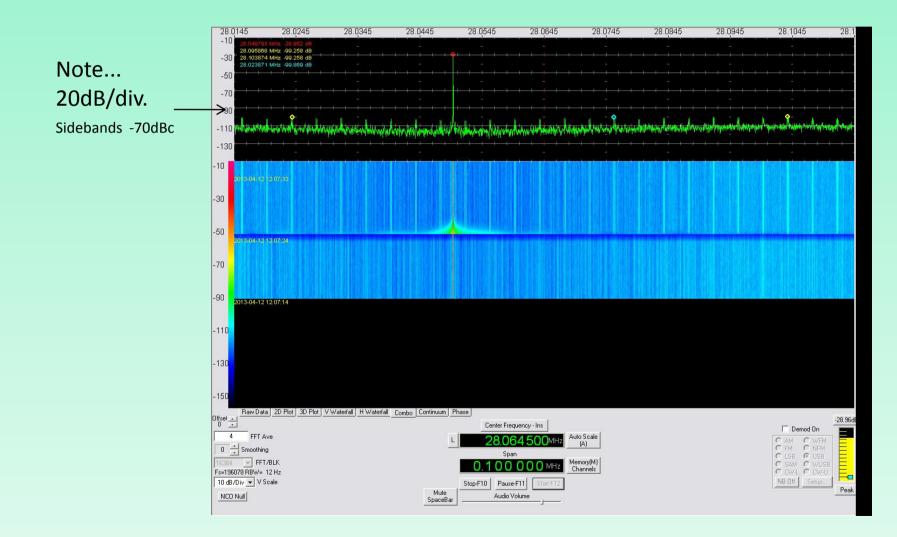
Type No	VCO	RF I/P Freq F	ract Res bits	O/P Divider	Vcc (Vph)	Availability
LMX2470	External	500 – 2600	22	-	2.5	RS £4.30
LMX2487	External	3000 - 7500	22	-	2.5 – 3.3	RS £5.30
LMX2541	Internal (6)	1990 – 4000	22	1-63	3.3 (3.3)	Digikey / Mouser ~\$10
ADF4150	External	500 - 5000	12	1,2,4,8,16	3.3 (5)	Farnell £5.40
ADF4150HV	External	500 - 3000	12	1,2,4,8,16	3.3 (30)	Farnell £6.20
ADF4156	External	500 - 6200	12	-	3.0 (5)	RS / Farnell £5.30
ADF4159	External	500 - 13000	25	-	1.8 & 3.0 (3)	
ADF4351	Internal	2200 – 4400 (full oct	ave) 12	2^N to 64	3.3 (3.3)	Digikey ~\$10
LMX2492	External	500 - 14000	24	-	3.3 (5)	Manuf, (Free samples)

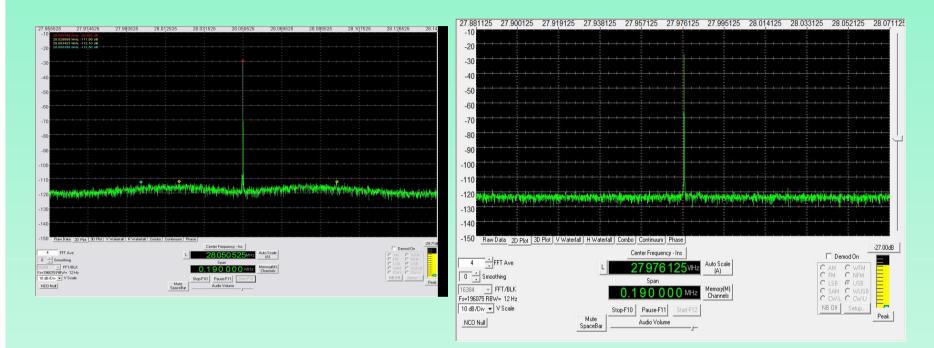
### A Few Spectral plots.

# Downconverted using overtone crystal oscillator LOs and SDR-IQ.

Good quality 10MHz TCXO is the reference – unknown phase noise

#### FractN Spurii at 432MHz Third and Fourth order Sigma-Delta



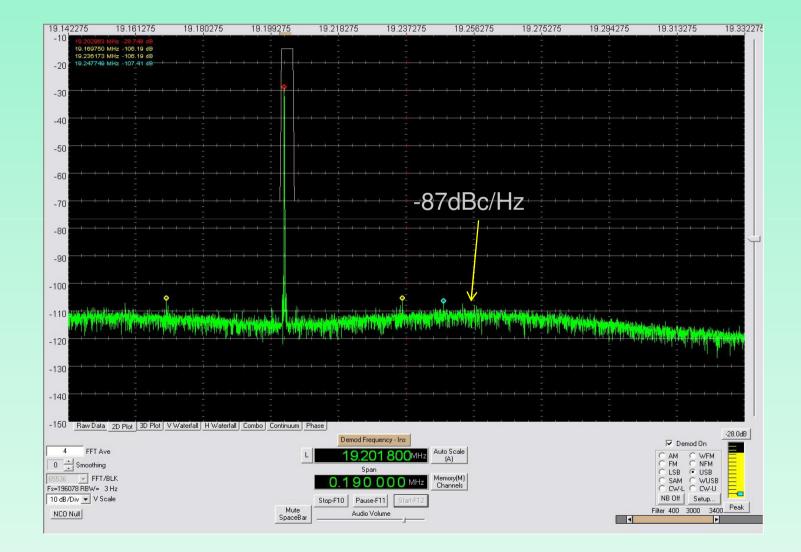


#### 432MHz (2.6GHz /6 OP Divider)

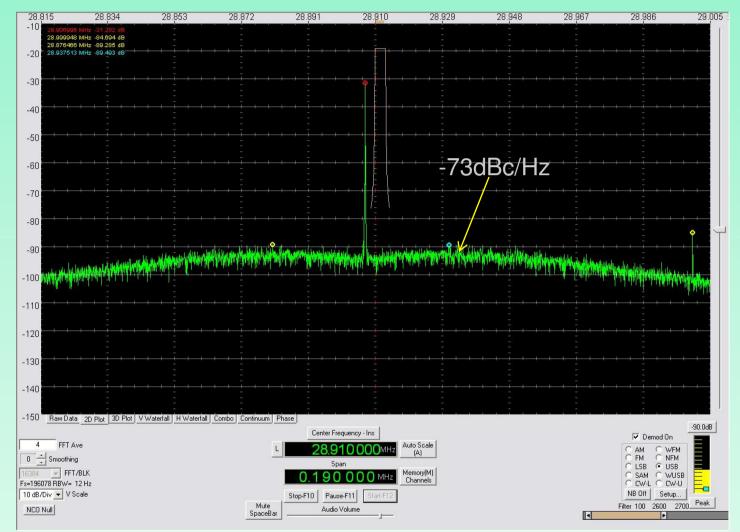
Compare with 432MHz from a crystal source

Both plots noise B/W = 12Hz. dBc/Hz is 11dB Lower

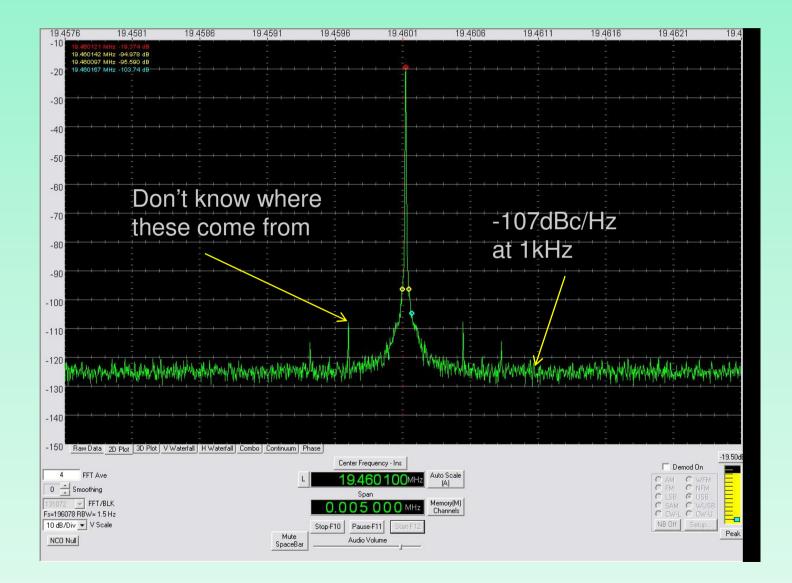
#### 1296.2MHz (5Hz step resolution) (Wide 80kHz loop bandwidth)



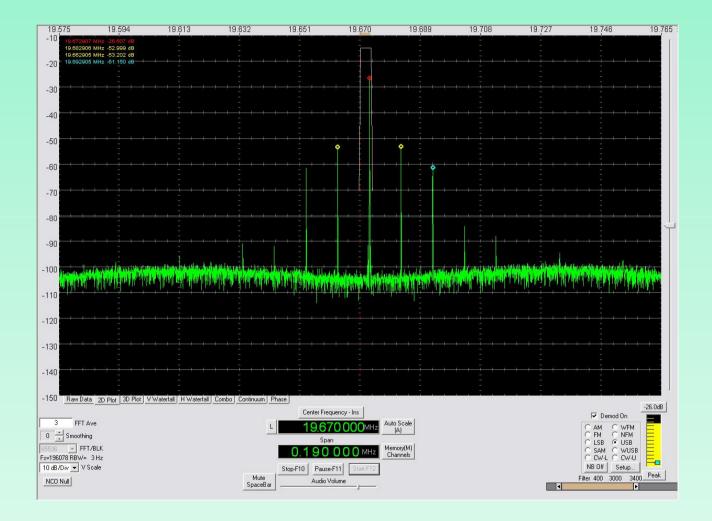
#### 5760MHz (LMX2541, 2.88GHz + Doubler) Loop filter 30kHz BW



#### 144MHz Close in (1.5Hz noise B/W)



#### Found by accident , an unfortunate choice of F/D ratio. 1296.67MHz 80kHz loop bandwidth



#### A convenient point to end if time is short.

If not...

A Few Practical Designs

### Practical Designs

- External VCO
  - LMX2470 up to 2.5GHz
  - LMX2487 up to 6GHz (both 22 bit D)

#### – ADF4150, 5GHz

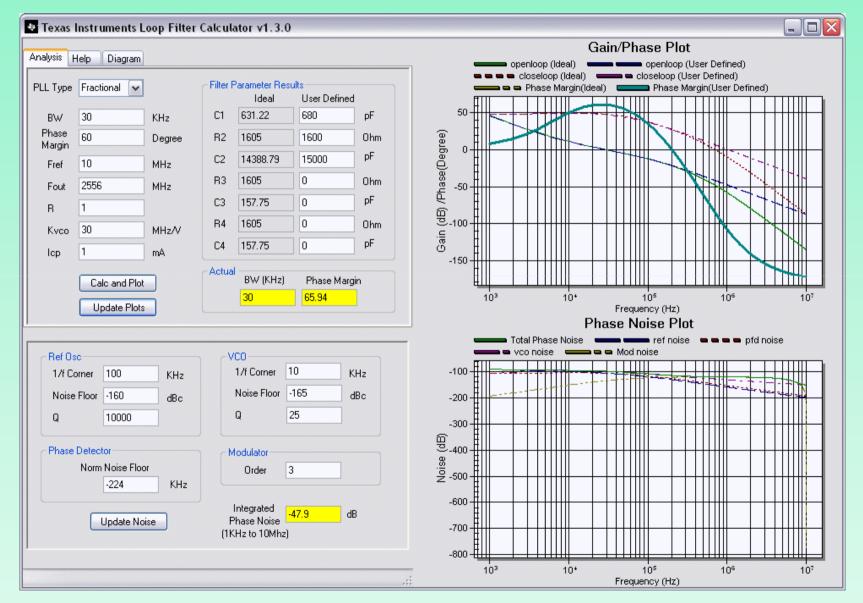
- 12 Bit Denominator (medium size steps)
- Output divider /2 , /4 , /8 , /16
- Lower VCO limit for all these is 500MHz

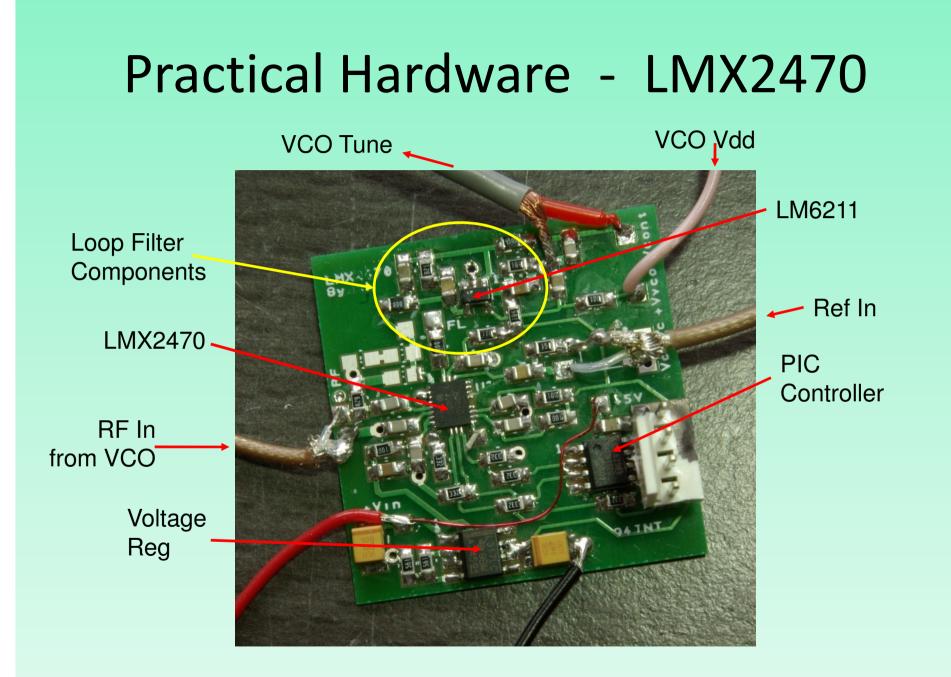
## **Tuning Voltage Limits**

- Charge Pump Vdd (or a rail not much higher)
  - 3 to 5V is typical
  - Except the ADF4150HV 30V Why did they
- External OPAMP / active filter
- LM6211 is designed for the job
  - 5 pin
  - 20MHz bandwidth
  - Rail to Rail

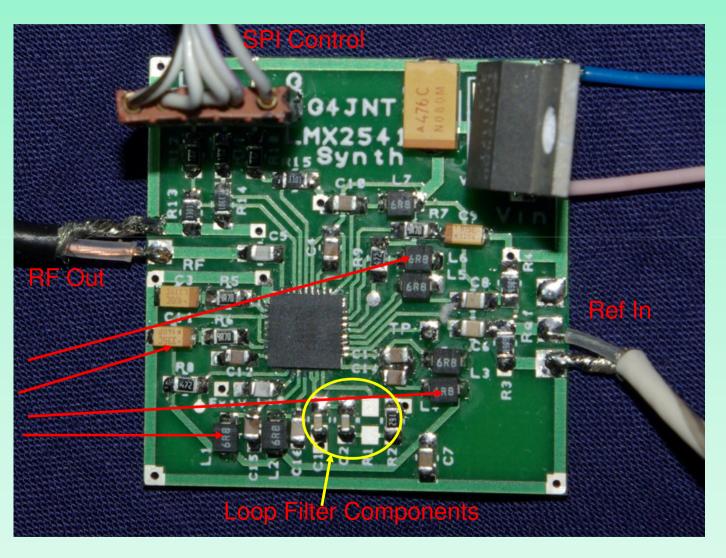
Why did they use the same type number ?

#### Loop Filter





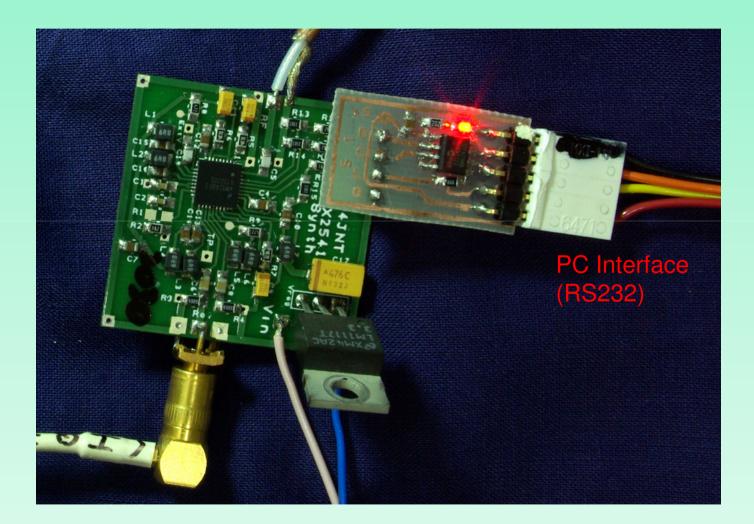
### It's easy with an Internal VCO!



Lots of Decoupling

But not much else

#### **RS232** Control



LMX2541 Control	G4JNT	
Output Freq MHz 432.431 Fix Fout		Fvco <b>3459.448 MHz</b> Version(s) allowed <b>541-3320E</b>
Ref Doubler Reference Input MHz	R divider 1	Fcomp <b>10000. kHz</b>
Divided O/P Resolution	MFSK Mode	N = 345 F = 1181000 D = 1250000
Update all registers Store to EE	<ul> <li>Charge Pump Gain</li> <li>20</li> <li>Strong Dither</li> <li>S-D Modulator</li> <li>Order 4</li> <li>Dig Lock Det</li> <li>Ext VC0</li> </ul>	12 Div Gain 12 VCO Gain 12 Out Term PIC Register Export
COM 1 7 00000017 12 0000001C 8 0111CE58 5 A0040005 3 01487303 1 04800011	13 000000CD 9 28001409 6 001F3326 4 FF4880A4 2 05312D02 0 05481590	

🛱 PIC Cut-and-paste	
$\begin{array}{c} ; \ Fout \ 432.431 MHz \ Resolution \ 1Hz \\ ; \ R = 1 \ D = 1250000 \ N = 345 \\ de \ 0x00, 0x00, 0x00, 0x00, 0x17 \ ; \ Reg7 \\ de \ 0x00, 0x00, 0x00, 0x00, 0xCD \ ; \ Reg13 \\ de \ 0x00, 0x00, 0x00, 0x1C \ ; \ Reg12 \\ de \ 0x28, 0x00, 0x14, 0x09 \ ; \ Reg9 \\ de \ 0x01, 0x11, 0xCE, 0x58 \ ; \ Reg8 \\ de \ 0x00, 0x14, 0x00, 0x05 \ ; \ Reg6 \\ de \ 0x40, 0x04, 0x00, 0x05 \ ; \ Reg5 \\ de \ 0x0F, 0x48, 0x80, 0xA4 \ ; \ Reg4 \\ de \ 0x01, 0x31, 0x2D, 0x02 \ ; \ Reg3 \\ de \ 0x05, 0x31, 0x2D, 0x02 \ ; \ Reg1 \\ de \ 0x05, 0x48, 0x15, 0x90 \ ; \ Reg0 \\ \end{array}$	F = 1181000

Software calculates register contents, then can send RS232 data and stores in the PIC for boot up.

