



6dB Better than CW

Weak Signal Modes on the Microwave
Bands

Andy Talbot G4JNT/G8IMR

Traditional CW

Is the weak signal mode used when all else
(especially SSB !) fails

Limited by

Noise – Proportional the bandwidth

Operators ability to decode it

Often need to repeat messages

Talkback / handshaking

Alphabet prone to errors if message is broken

J (--- --- ---) = E E M (- - - -)

A few Values

- Ear / Brain combination is surprisingly good at detecting tones buried in noise
 - SSB voice needs $\sim 3\text{kHz}$ for full readability
 - We detect it as if the bandwidth were only a few tens of Hz
 - And especially tones at the right frequency, in something like 20 – 100Hz noise bandwidth
- This often gives a false impression how good a signal really is.

Some Audio Generated by Maths!

Sampling Rate 8000 Hz --- 16 Bit sampling, noise RMS = -10dB
 S/N Ref BW 2500 Hz -5dB = $32767 / \text{SQRT}(10) = 10361$
 Audio BW 100 Hz -19dB 99% probability of no clipping

Signal/Noise

Source File	Amplitude dB wrt FSD	Dest File	Noise dB FSD	2.5kHz Norm.	100Hz Audio
CWMSGM20 	-20	..\dsp\weakcw01.wav 	-10	-5	9
CWMSGM25 	-25	..\dsp\weakcw02.wav 	-10	-10	4
CWMSGM30 	-30	..\dsp\weakcw03.wav 	-10	-15	-1
CWMSGM35 	-35	..\dsp\weakcw04.wav 	-10	-20	-6 CW Limit ?
CWMSGM40 	-40	..\dsp\weakcw05.wav 	-10	-25	-11

The Limit for CW

- At best, around 10dB S/N in 30Hz for *easy copy* CW. No repeats
 - Several dB lower for detecting
- Assume 18WPM at this level.
- A Word has 5 chars, 4.5 bits / char (plain text) = 6 or 7 Bits /second equiv data rate
- Repeat the message, gives $\sim 2 - 3$ Bits/s
 - This is manual *Forward Error Correction*

Compare simple data modes

- RTTY 30 Bits/s (for 50 baud)
 - Needs $>15\text{dB}$ S/N in 100Hz (around the two tones)
 - Very inefficient
- PSK31 $>10\text{dB}$ S/N in 31Hz
 - About the same as CW with a good operator
- Both of these are error prone, so repeats are needed
 - Reduces overall data rate

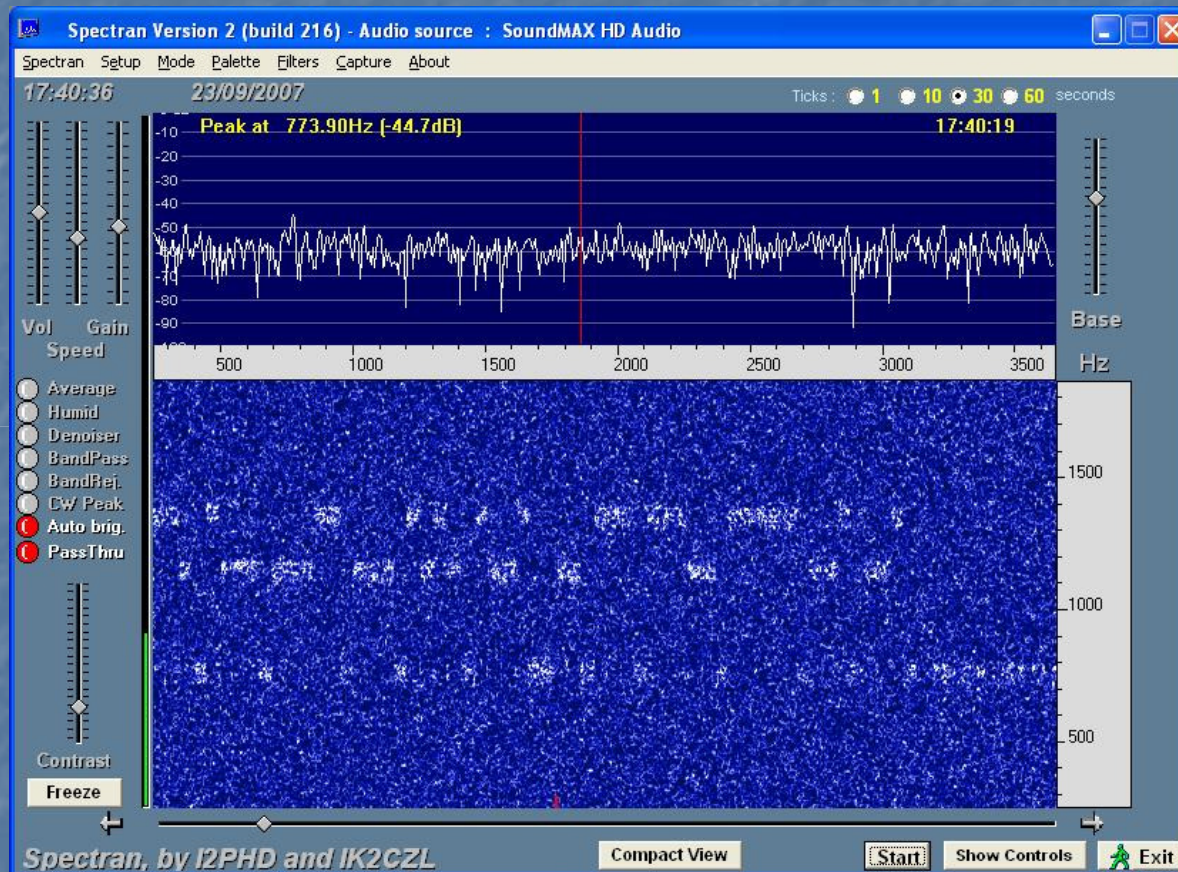
And what about Microwave Bands?

- Rain Scatter
 - Spreads the signal over 100 – 300Hz
 - CW survives this quite well, (and RS is often strong)
 - DFCW with spectral display works reasonably well
- Scattering / breakup / troposcatter / multipath
 - Kills CW
 - Kills most other data modes

Can we go Narrower

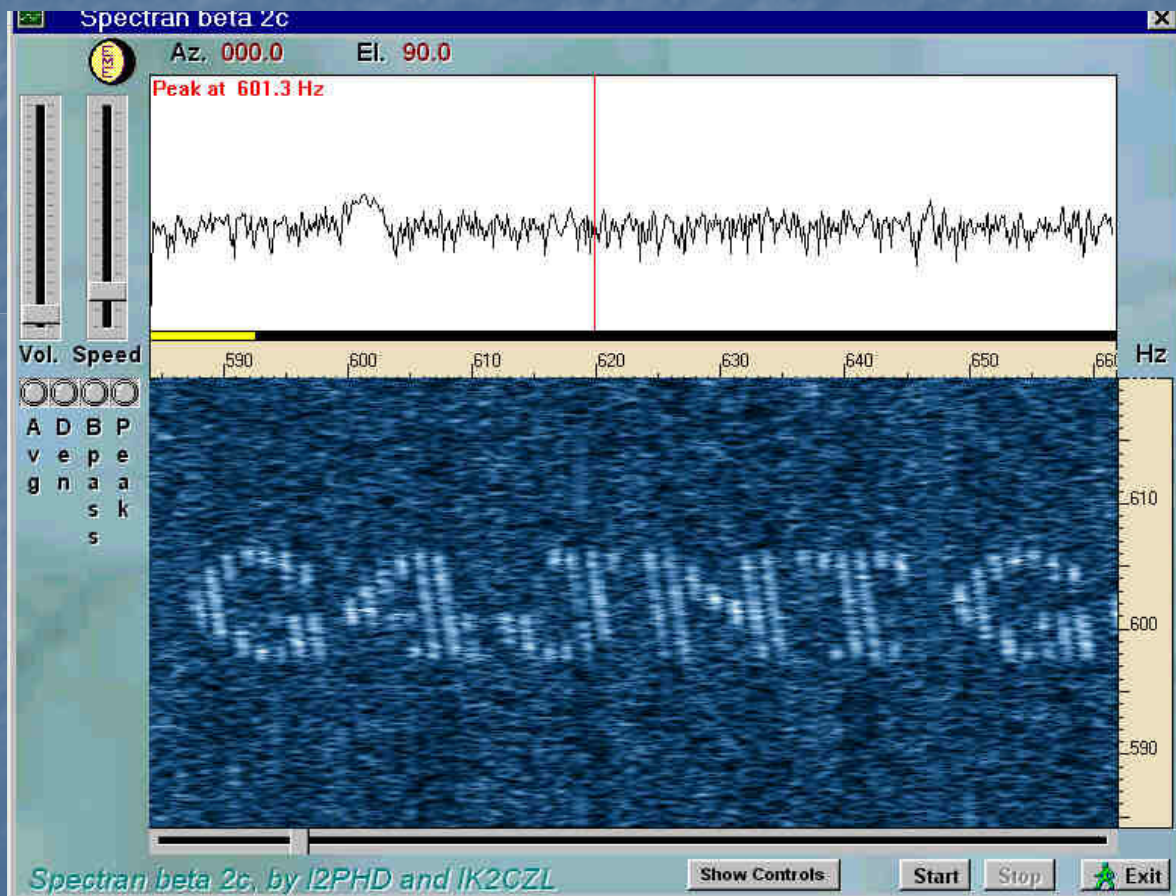
- Yes - Lower BW , less noise, increases S/N
- But the message takes proportionately longer to send
- Spreading could be a problem
- Need machine assistance
 - QRSS, Visual CW, DFCW, Hell
- `About the same signalling efficiency as CW`

Visual CW / DFCW / DFCWi



Advantages for very slow data in narrow band
– about same as aural CW when scaled for speed/BWidth.

SMT Hell

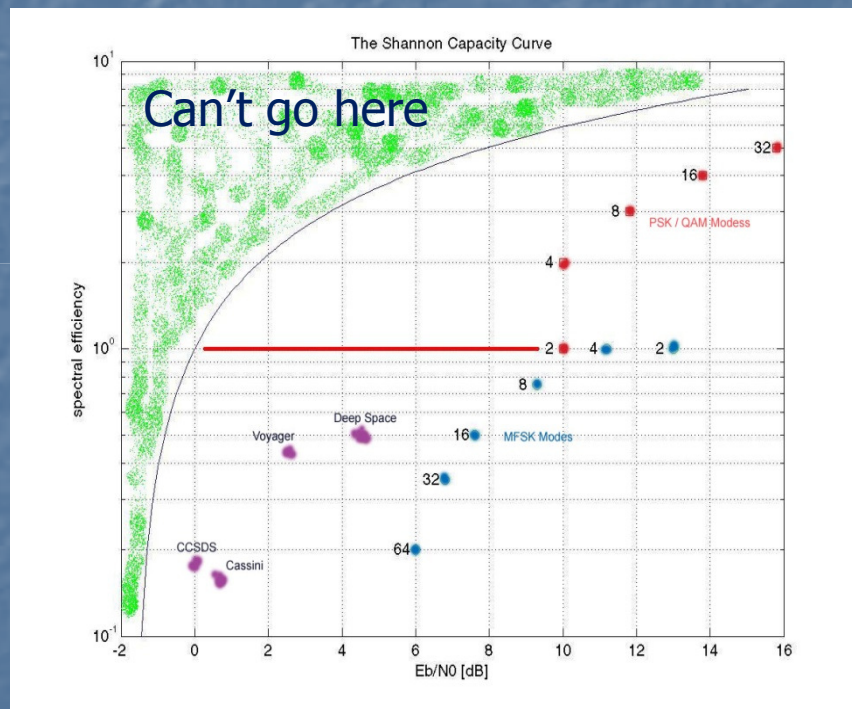


Can We Do Better

- Yes – but there is a theoretical limit
- Shannon suggests two routes
 - Improve the signalling efficiency with better modulation (move left)
 - Not much scope with modern modulations
 - Increase the bandwidth (move up)
 - This appears counter-intuitive BUT-----
 - The maths sez so

The Shannon Curve

Was derived from basic Physics / Maths / Info theory and is NOT an experimental result. It is a TARGET.



- Horizontal – Normalised Signal/Noise, energy / bit
 - Move left, Lower Tx power, or increase noise
- Vertical – Spectral Efficiency, Bits/s/Hz.
 - Move up, increase bandwidth for same capacity
- Red – 2 ... 32 PSK/QAM
- Blue – 2 ... 64 MFSK
- Purple - Heavily coded deep space
- Red Line – “10dB from Shannon”

<http://marconig.wordpress.com/2007/07/03/the-shannon-capacity-curve/>

Bandwidth Expansion

- Commercially / military use spread spectrum
 - WLAN, Bluetooth, Wi-Fi.
 - All improve signalling efficiency by spreading the signal over a wide bandwidth to counter interference / multipath
- Not too easy on the Am. Bands as we nearly always want to keep within the 3kHz SSB bandwidth

Another Way

- Heavy Error Correction
 - Often not thought-of as a bandwidth spreading
 - We already see it in normal operation –
 - Repeat the information many times
 - Slowing the data rate and keeping the same modulation format *is equivalent to widening the bandwidth*
 - It's the ratio of Data Rate / Bandwidth that matters

Source Coding

- First Get rid of redundant information (WSJT Style)
 - Compress callsigns using their known structure
 - Char-Char-Number-Letter-Letter-Letter
 - Letter = A-Z or [space]. Char = Letter or 0-9
(but note the 2nd Char cannot be a [space])
 - Compresses to $37*36*10*27*27*27 = 262\text{Meg}$
 - Which can be represented by 28 Bits
 - (RTTY needs at least 35 bits, could be more depending on letter/figure shift)
- Locator (4 digit) $18*18*10*10 = 32400$ (15 Bits)
 - 6 Digit Loc 25 bits

Further Source Coding

- Assume 4 Million Radio Ams in the world (*we wish!*)
 - Use a codebook to store the callsign of everyone, then just transmit the reference number
 - Only needs 22 bits
 - This is contentious lets not go there !
- Reports and acknowledgements need only a few bits in reality
 - But this also sparks controversy
- With the natural redundancy removed, any random data message begins to look valid
 - Acknowledged 'problem' with source coding

An Aside....

- Morse is a classic example of source coding
 - Most common letters use less data bits than less popular ones
 - Same problem of one symbol being corrupted to another
 - eg. T = E E
 - Bleeps from continuity tester can spell messages

Modulation

- On-Off, or Amplitude Shift Keying is not good.
 - It must waste 3dB
- PSK is theoretically the best (multiplication by 1 or -1)
 - Maintains high duty/cycle
 - Coherency needs frequency / phase lock
 - Which can be destroyed by propagation anomalies
 - Non-linear processing for recovery throws away many of the advantages of coherent reception
 - Unless bandwidth is unimportant, needs linear transmitters
- Which leaves good old fashioned, well established FSK

Multi FSK

- Use several Tones
 - Extend these over more than the anticipated spread
 - 10's of Hz for V/UHF.
 - 100's of Hz for microwave
 - All well within the 3kHz SSB bandwidth.
- 4 tones give 2 bits per symbol
 - F0 = '00', F1 = '01', F2 = '11', F3 = '10' WSPR / JT4
- 64 tones 6 bits per symbol
 - F0 = '000000', F7 = '000111', F26 = '011010', F63 = '111111' JT65
 - We've increased our data rate at the expense of decoding complexity – that's no problem these days

Error Correction

- Now make good use of our increased capacity / data rate
 - Could just repeat the message several times and compare each, looking for errors in each bit.
 - Three repeats allows error correction
 - Two repeats allows detection – may be enough if talkback allows a repeat request
 - Interleave the repeats to counter burst errors
 - But we can do a lot better
 - and its very mathematical

Error Correction Techniques

■ Hamming Distance

- Add enough extra parity bits so the new alphabet has a certain number of bits different between each block. Then compare each received one and look for the most probable.
 - Example is 4 bits with 3 more parity
 - Allows 1 error in a total of 7 to be corrected
 - 2 errors can be detected
 - Simple schemes are decoded using lookup tables

■ Block coding

- More efficient longer-word schemes are in widespread use
- Reed-Solomon, BCH
 - But the maths processing is NOT NICE
 - Galois Fields, Dividing Polynomials

Error Correction Techniques continued

- Convolutional Coding
 - Continuously spread each source over several bits of the output. Adding more for correction – eg x2 or x3
 - Continuously look for what was *most likely to have been sent* in order to generate what has *actually been received*.
 - Soft decision decoding looks at probability a received symbol is good, bad or indifferent
 - The Viterbi decoding algorithm
 - Searches back through received symbols in a trellis, looking for the most likely data that could have generated it
 - Processor intensive, adds a delay.

Another Aside

A few state-of-the art codes

- Taken from
 - <http://marconig.wordpress.com/2007/07/03/the-shannon-capacity-curve/>
- These are for BPSK with the coding used with several deep space (interplanetary) spacecraft:
- $r=1/2$ $k=7$ convolutional: E_b/N_0 4.5 dB, eff 0.5 bps/Hz
- Voyager (RS+ $r=1/2, k=7$): E_b/N_0 2.4 dB, eff 0.437 bps/Hz
- Cassini (RS+ $r=1/6$ $k=15$): E_b/N_0 0.6 dB, eff 0.146 bps/Hz
- CCSDS $r=1/6$ turbo large block: E_b/N_0 0.0 dB 0.167 bps/Hz
- Not much scope for further improvement

Timing and Frequency Errors

- Need knowledge of frequency / tuning error and timing
 - Use UTC based protocol to limit search requirements
 - Identify Start of message timing
 - To be able to identify the right symbols
 - Can't afford to spend a lot of time searching
 - Typical few seconds for PC clock errors, bit more for EME delays
 - Frequency get within a tone bandwidth for MFSK schemes.
 - Send synchronisation Sequence
 - Unique pattern to search for that won't appear anywhere in the message. Can give frequency and time.

WSJT Examples

■ JT65

- 72 source Bits - 2 compressed callsigns + one 4-digit Locator OR 13 chars of plain text.
- Block coding (Reed Solomon) expanded to 126 symbols of 64 tones (6 bits / symbol) ,and one more for sync , Pseudo Random interspersed.
- Effectively expands a 72 bit message to an effective 441 bits
- Big Sync overhead – 50% of the message time
- Three tone spacings, 2.7, 5.4 and 10.8Hz

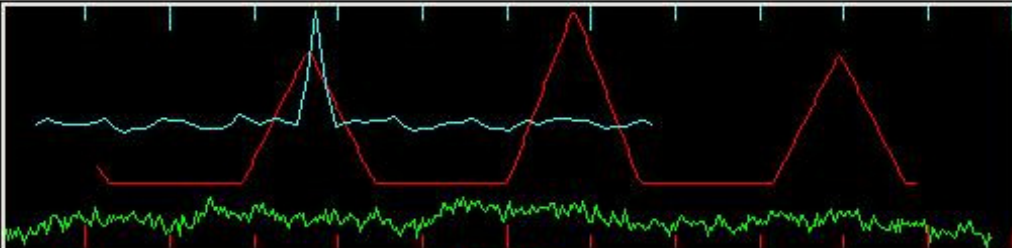
JT4 a-g and WSPR

- Both similar *coding* schemes
 - Four tones carrying two bits per symbol,
 - One bit is sync sent as a pseudo-random code
 - The other is a data bit
 - JT4 same message as JT65,
 - 72 bits expanded to 207 in a convolutional encoder
 - Sent in 48 seconds at 4.375 symbols/s
 - Tone spacing user selected from 4.4 to 315Hz
 - WSPR Different Message, new data structure
 - 50 bits expanded to 162 in a convolutional encoder
 - Sent in 110 seconds at 1.46 symbols / second
 - Tone spacing 1.46Hz

WSJT User Screen

WSJT 7 by K1JT

File Setup View Mode Decode Save Band Help



Moon
 Az: 113.17
 El: 21.86
 Dop: 10412
 Dgrd: -1.8

6.0 Time (s) Mon_091030_172600

FileID	Sync	dB	DT	DF	W				
172100	0	-3	1.7	-22	0				
172200	24	-5	1.8	74	0 *	GB3SCS	I080UU	1	0
172300	0	-3	5.4	94	6				
172400	23	-2	1.7	98	35 *	GB3SCS	I080UU	1	0
172500	0	-3	1.7	88	43				
172600	24	-2	1.7	79	84 *	GB3SCS	I080UU	1	0

Log QSO
Stop
Monitor
Save
Decode
Erase
Clear Avg
Include
Exclude
TxStop

To radio: GB3SCS

Grid: io80uu

Az: 267 70 km

2009 Oct 30
17:27:15

Sync 1 Zap

Clip 0 NB

Tol 200 Freeze

Defaults AFC

Dsec 0.0 Shift 0.0

Tx First GB3SCS G4JNT IO90 Tx1

25 Rpt GB3SCS G4JNT IO90 000 Tx2

Sh Msg RO Tx3

TxDF = 0 RRR Tx4

GenStdMsgs 73 Tx5

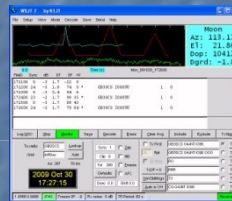
Auto is Off CQ G4JNT IO90 Tx6

1.0068 0.9998
JT4G
Freeze DF: -9
Rx noise: 0 dB
TR Period: 60 s
Receiving

Using WSJT

- Setup Box
 - Callsign, Locator, Com Port for Tx control
 - Make Sure sampling rate calibration is OK
 - (Only done once per PC – unless using .WAV files). Look at Self Check value. Enter into Setup, Options
- Set The right Mode (easily forgotton!)
- Set PC Clock
 - Dsec Box for fine tuning – aim for less than a second or two error from UTC
- Adjust Audio Levels
 - Need Rx or Monitor to be running

Run WSJT.....



Load in .WAV files from GB3SCX and GB3SCS
Set Rate in to 1.0068 (Saved on a different machine)

Replay .WAV files and use mic to loop round
Set Options Rate-in back to to 0.9797 - check value.
(Although they were recorded on another machine at 11100Hz, check exact value!)
Use Monitor mode and start VLC replay 2 seconds early

Where to hear WSJT Signals

- Off the Moon , JT65A, B, C
- GB3SCX 10368.905MHz JT4G
- GB3SCS 2320.905MHz JT4G
 - Tune so USB carrier is 800Hz below
- GB3VHF 144.43MHz JT65B
 - Tune 1500Hz low, USB carrier 144.4285MHz
- GB3RAL 40.05/50.05/60.05/70.05 JT65B
 - Tune USB carrier Xx.0485MHz
- HF Bands JT65A, JT4A+, WSPR

The Shannon Capacity Curve

